

# Multibeam Analog Beamformer Design for Monostatic ISAC under Self-Interference

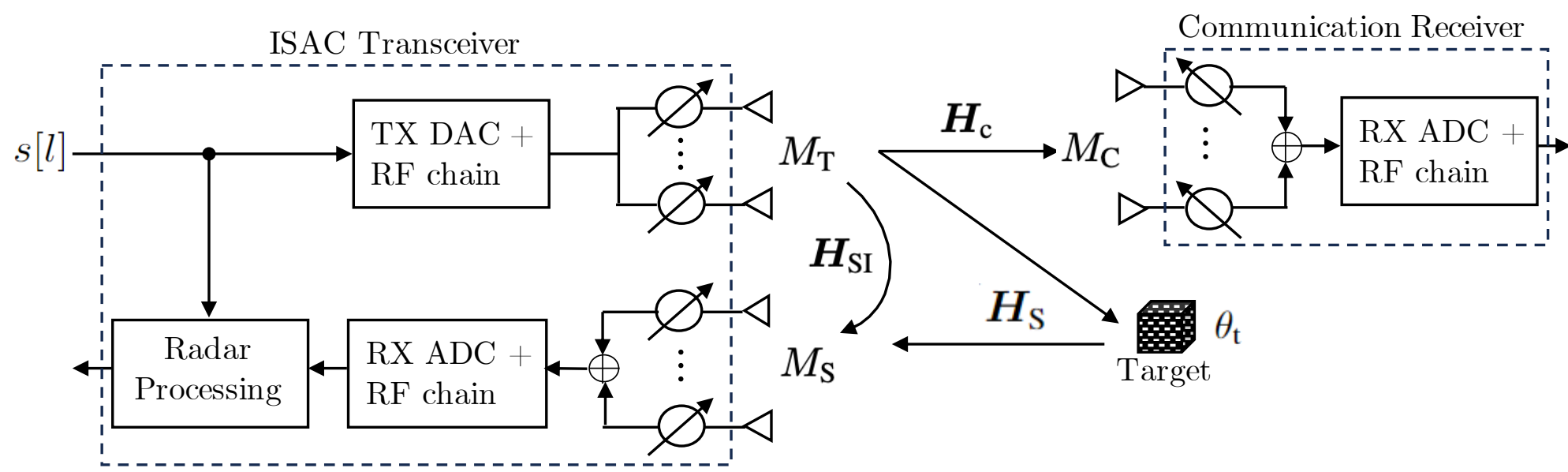
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## Summary

- This work explores multibeam (dual-functional) fully analog beamforming for joint communication and monostatic sensing under self-interference (SI).
- We provide a semi-analytic optimal solution under total power constraints and its adaptation to constant-modulus (CM) analog beamformers.

## Problem Setting



- Assume CM phased arrays: each antenna is connected to a single phase shifter with no individual gain control.
- The ISAC transceiver sends a data stream  $\{s[l]\}$  of complex zero-mean, unit-variance symbols. The TX signal reads:

$$x[l] = f s[l] \in \mathbb{C}^{M_T}$$

- The observation at the co-located radar is given by

$$r[l] = \mathbf{w}_s^H \mathbf{H}_S \mathbf{f} s[l] + \underbrace{\mathbf{w}_s^H \mathbf{H}_{SI} \mathbf{f} s[l]}_{\text{Post-combining SI}} + \mathbf{w}_s^H \mathbf{n}_s[l]$$

- Post-combining SI due to weak isolation between TX and sensing arrays and  $\ell[\{s[l]\}] >$  target's echo RTT.
- SI degrades radar performance and saturates RF front-end.
- Conventional approach:

$$\min_{\mathbf{f}, \mathbf{w}_s} \mathbb{E} \{ \mathbf{w}_s^H \mathbf{H}_{SI} \mathbf{f} s[l] \}$$

→ Does not protect RF components preceding  $\mathbf{w}_s$

- Solution:** Control *pre-combining* SI

$$P_{SI} \triangleq \mathbb{E} \{ \|\mathbf{H}_{SI} \mathbf{f} s[l]\|^2 \} = \mathbf{f}^H \mathbf{H}_{SI}^H \mathbf{H}_{SI} \mathbf{f}$$

**Objective:**

$$\begin{aligned} \max_{\mathbf{f} \in \mathbb{C}^{M_T}} \quad & G_{\text{tx,comm}} \triangleq \mathbf{f}^H \mathbf{H}_c^H \mathbf{H}_c \mathbf{f} \\ \text{s.t.} \quad & G_{\text{tx,sen}}(\theta_t) \triangleq |\mathbf{f}^H \mathbf{a}_T(\theta_t)|^2 \geq \tau^2 \\ & P_{SI} \leq \eta^2 \\ & |f_i| = 1, i \in \{1, \dots, M_T\} \end{aligned}$$

## Design under Total Power Constraints

- To gain insight: relax CM constraints

$$|f_i| = 1, i \in \{1, \dots, M_T\} \rightarrow \|\mathbf{f}\|^2 = M_T$$

- The design problem becomes

$$\begin{aligned} \max_{\mathbf{f} \in \mathbb{C}^{M_T}} \quad & G_{\text{tx,comm}} \\ \text{s.t.} \quad & (a) G_{\text{tx,sen}}(\theta_t) \geq \tau^2, (b) P_{SI} \leq \eta^2, (c) \|\mathbf{f}\|^2 = M_T \\ & \rightarrow \text{QCQP with 3 constraints} \end{aligned}$$

- First order optimality condition:

$$\mathbf{f}_* = \sqrt{M_T} \mathcal{D} [\mathbf{H}_c^H \mathbf{H}_c + \mu_* \mathbf{a}_T(\theta_t) \mathbf{a}_T^H(\theta_t) - \gamma_* \mathbf{H}_{SI}^H \mathbf{H}_{SI}]$$

→  $\mathcal{D}[\mathbf{X}]$ : dominant unit-norm eigenvector of  $\mathbf{X}$

→  $\mu_*, \gamma_*$ : optimal Lagrange multipliers for (a) and (b)

- Procedure to attempt to solve the problem:

- If the solution obtained by relaxing constraints (a) and (b) is feasible:

$$\mathbf{f}_* = \sqrt{M_T} \mathcal{D} [\mathbf{H}_c^H \mathbf{H}_c]$$

Otherwise, proceed to Step 2.

- Relaxing either constraint (a) or (b):

- Relax constraint (b) and solve

$$\begin{aligned} \mathbf{f}_{(a)} &= \arg \max_{\mathbf{f} \in \mathbb{C}^{M_T}} G_{\text{tx,comm}} \\ \text{s.t.} \quad & G_{\text{tx,sen}} = \tau^2, \|\mathbf{f}\|^2 = M_T \end{aligned}$$

- Relax constraint (a) and solve

$$\begin{aligned} \mathbf{f}_{(b)} &= \arg \max_{\mathbf{f} \in \mathbb{C}^{M_T}} G_{\text{tx,comm}} \\ \text{s.t.} \quad & P_{SI} = \eta^2, \|\mathbf{f}\|^2 = M_T \end{aligned}$$

- Let  $\alpha \triangleq \mathbf{f}_{(a)}^H \mathbf{H}_{SI}^H \mathbf{H}_{SI} \mathbf{f}_{(a)}$  and  $\beta \triangleq \mathbf{f}_{(b)}^H \mathbf{a}_T(\theta_t) \mathbf{a}_T^H(\theta_t) \mathbf{f}_{(b)}$ . Then,

$$\mathbf{f}_* = \begin{cases} \mathbf{f}_{(a)} & \text{if } \alpha \leq \eta^2 \text{ and } \beta \leq \tau^2 \\ \mathbf{f}_{(b)} & \text{if } \alpha \geq \eta^2 \text{ and } \beta \geq \tau^2 \\ \arg \max_{\{\mathbf{f}_{(a)}, \mathbf{f}_{(b)}\}} G_{\text{tx,comm}} & \text{if } \alpha \leq \eta^2 \text{ and } \beta \geq \tau^2 \end{cases}$$

If  $\alpha \geq \eta^2$  and  $\beta \leq \tau^2$ , proceed to Step 3.

- The optimal  $\mu_*$  and  $\gamma_*$  must simultaneously satisfy  $\mathbf{f}_*^H \mathbf{a}_T(\theta_t) \mathbf{a}_T^H(\theta_t) \mathbf{f}_* = \tau^2$  and  $\mathbf{f}_*^H \mathbf{H}_{SI}^H \mathbf{H}_{SI} \mathbf{f}_* = \eta^2$ .

→ Newton-Raphson method

## Design under CM constraints

- Restoring the original CM constraints, the problem reads

$$\begin{aligned} \max_{\mathbf{f} \in \mathbb{C}^{M_T}} \quad & G_{\text{tx,comm}} \\ \text{s.t.} \quad & (a) G_{\text{tx,sen}}(\theta_t) \geq \tau^2, (b) P_{SI} \leq \eta^2, (c) |f_i| = 1, \forall i \end{aligned}$$

- Some options to enforce CM constraints:

- Project TPC solution onto the set of CM vectors
- Replace  $|f_i| = 1$  by  $|f_i|^2 = 1$  and solve a QCQP with  $M_T + 2$  constraints via SDR

- Proposal:** Modify the TPC procedure to enforce CM constraints at each iteration:

$$\mathcal{P}_{\mathbb{V}^n} \{\mathbf{x}\} = [x_1/|x_1|, \dots, x_n/|x_n|]^T$$

→ At Step 2, find  $\mathbf{f}_{(a)}$  and  $\mathbf{f}_{(b)}$  with:

### CM Bisection Search( $\mathbf{X}, \mathbf{Y}, \beta, \epsilon, \omega_{\min}, \omega_{\max}$ )

- Repeat

- $\omega \leftarrow (\omega_{\min} + \omega_{\max})/2$
- $\mathbf{f} \leftarrow \mathcal{P}_{\mathbb{V}^{M_T}} \{\mathcal{D}[(1 - \omega)\mathbf{X} + \omega\mathbf{Y}]\}$
- if  $\mathbf{f}^H \mathbf{Y} \mathbf{f} > \beta$ , then  $\omega_{\max} \leftarrow \omega$
- else  $\omega_{\min} \leftarrow \omega$

- Until  $|\mathbf{f}^H \mathbf{Y} \mathbf{f} - \beta| \leq \epsilon$

→ At Step 3, find  $\mathbf{f}_*$  with:

### CM Newton-Raphson( $\mathbf{X}, \mathbf{Y}, \mathbf{Z}, \beta_1, \beta_2, \epsilon, \alpha_0$ )

- $\alpha \leftarrow \alpha_0$
- $\mathbf{f} \leftarrow \mathcal{P}_{\mathbb{V}^{M_T}} \{\mathcal{D}[\mathbf{X} + \alpha[1]\mathbf{Y} - \alpha[2]\mathbf{Z}]\}$
- Repeat

- $\mathbf{u} \leftarrow \mathcal{P}_{\mathbb{V}^{M_T}} \{\mathcal{D}[\mathbf{X} + \alpha[1]\mathbf{Y} - \alpha[2]\mathbf{Z}]\}$
- Set

$$\mathbf{q}(\alpha) \leftarrow \begin{bmatrix} \mathbf{u}^H \mathbf{Y} \mathbf{u} - \beta_1 \\ \mathbf{u}^H \mathbf{Z} \mathbf{u} - \beta_2 \end{bmatrix}.$$

- Compute the Jacobian matrix

$$\mathbf{J}_q(\alpha) \leftarrow \begin{bmatrix} \frac{\partial}{\partial \alpha[1]} q_1(\alpha) & \frac{\partial}{\partial \alpha[2]} q_1(\alpha) \\ \frac{\partial}{\partial \alpha[1]} q_2(\alpha) & \frac{\partial}{\partial \alpha[2]} q_2(\alpha) \end{bmatrix}$$

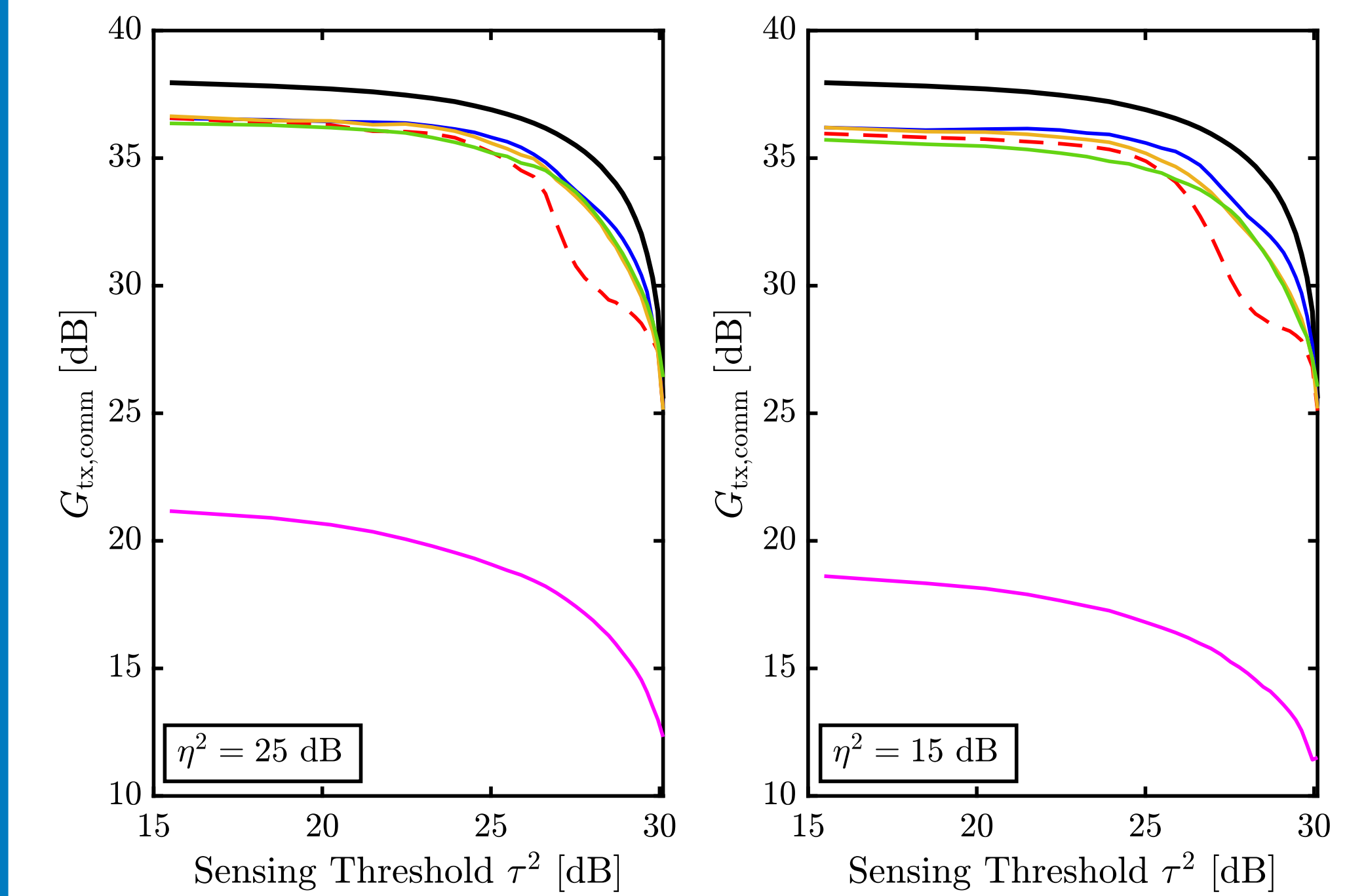
- Update  $\alpha \leftarrow \alpha - \mathbf{J}_q^{-1}(\alpha) \mathbf{q}(\alpha)$

- $\mathbf{f} \leftarrow \mathcal{P}_{\mathbb{V}^{M_T}} \{\mathcal{D}[\mathbf{X} + \alpha[1]\mathbf{Y} - \alpha[2]\mathbf{Z}]\}$

- Until  $|\mathbf{f}^H \mathbf{Y} \mathbf{f} - \beta_1| \leq \epsilon$  &  $|\mathbf{f}^H \mathbf{Z} \mathbf{f} - \beta_2| \leq \epsilon$

## Numerical Example

- Proposed with CM (blue)
- Communication-Sensing Upperbound: neglect SI (black)
- Benchmarks:
  - SDR with CM (dashed red)
  - Convex Combination with CM (magenta)
  - Proposed with TPC + Final Projection (orange)
  - SDR with TPC + Final Projection (green)



- SDR with CM
  - Unable to find rank-1 solutions in 40% of the trials
- Proposed with TPC + Final Projection
  - Prob. exceeding SI limit: 20% (left) and 49% (right)
- SDR with TPC + Final Projection
  - Almost surely violates SI constraint

## References & Acknowledgements

- J. Borrás and R. López-Valcarce, "Multibeam analog beamformer design for monostatic isac under self-interference," in *2026 IEEE International Conference on Acoustics, Speech and Signal Processing (ICASSP)*, 2026.

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