

Summary

• What we propose:

- A channel-aware semi-unitary precoder that minimizes OBR over $\mathcal{B}$.
- The spectral front-end is obtained from the eigenvectors associated with the smallest eigenvalues of the OBR matrix Φ .
- Right-unitary invariance embeds CSIT without changing the transmit PSD/OBR.
- SVD- and GMD-based rotations enable low-complexity ZF and SIC receivers, respectively.

System Overview

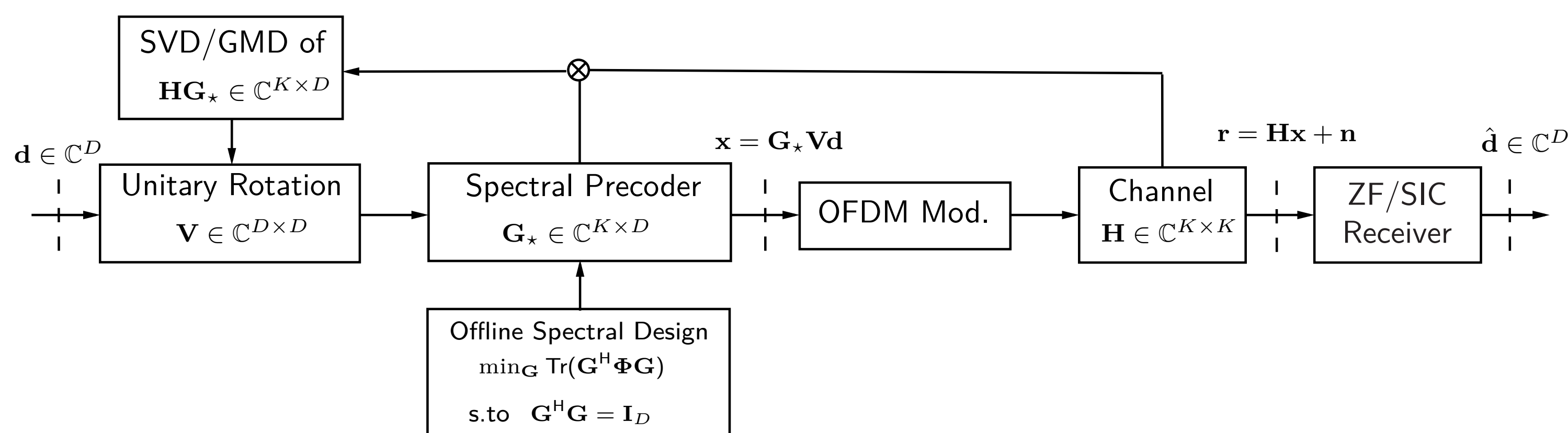


Fig.1 Proposed Channel-Aware Spectral Precoding Chain.

• Spectral precoding:

- Map $D \leq K$ symbols $\mathbf{d} \in \mathbb{A}^D$ onto the K active tones: $\mathbf{x} = \mathbf{G}\mathbf{d} \in \mathbb{C}^K$
- $\mathbf{G} \in \mathbb{C}^{K \times D}$: with redundancy $R = K - D$, coding rate $\lambda = D/K \leq 1$.

• PSD structure:

- Under $\mathbb{E}[\mathbf{d}_m \mathbf{d}_{m-\ell}^H] = \delta[\ell] \mathbf{I}_D$, the continuous-time PSD has the form [1]:

$$P_s(f) \propto \phi^H(f) \mathbf{G} \mathbf{G}^H \phi(f).$$

- $\phi(f) \in \mathbb{C}^K$: stacks the spectral shapes of the active subcarriers.
- PSD depends on $\mathbf{G}$ through $\mathbf{G} \mathbf{G}^H$ (invariant to right-unitary rotations).

• Channel:

- Under standard CP-OFDM, the post-FFT channel is diagonal:

$$\mathbf{H} = \text{diag}\{H_c[k_1], \dots, H_c[k_K]\} \in \mathbb{C}^{K \times K}.$$

- $H_c[k]$: N -point DFT sample of the channel impulse response at k .

• Received signal:

- The received vector (after CP removal and N -point FFT):

$$\mathbf{r} = \mathbf{H}\mathbf{G}\mathbf{d} + \mathbf{n}, \quad \mathbf{n} \sim \mathcal{CN}(\mathbf{0}, \sigma^2 \mathbf{I}_K).$$

Optimization Framework

• Problem statement:

- Define the weighted transmit power over the protected band $\mathcal{B}$

$$P_W \triangleq \int_{-\infty}^{\infty} W(f) P_s(f) df = \text{Tr}(\mathbf{G}^H \Phi \mathbf{G}), \quad \Phi \succeq \mathbf{0}.$$

- $\Phi \in \mathbb{C}^{K \times K}$ captures leakage weighted by $W(f) \geq 0$ in $\mathcal{B}$.
- To avoid the trivial solution $\mathbf{G} = \mathbf{0}$, enforce a semi-unitary constraint:

$$\min_{\mathbf{G}} \text{Tr}(\mathbf{G}^H \Phi \mathbf{G}) \quad \text{s.t.} \quad \mathbf{G}^H \mathbf{G} = \mathbf{I}_D$$

• Closed-form solution:

- Consider the eigendecomposition

$$\Phi = \mathbf{Z} \text{diag}\{\nu_1, \dots, \nu_K\} \mathbf{Z}^H, \quad \nu_1 \geq \nu_2 \geq \dots \geq \nu_K \geq 0$$

where $\mathbf{Z} = [\mathbf{z}_1, \dots, \mathbf{z}_K]$ is unitary.

- One optimal semi-unitary precoder is

$$\mathbf{G}_* = [\mathbf{z}_{K-D+1}, \mathbf{z}_{K-D+2}, \dots, \mathbf{z}_K]$$

- Hence, $\mathbf{G}_*$ spans the eigenspace associated with the D smallest eigenvalues of Φ .
- $\mathbf{G}_*$ is unique up to a right-unitary rotation: $\mathbf{G}_* \mathbf{V}$ is spectrally equivalent, $\mathbf{V} \in \mathbb{C}^{D \times D}$ is free for CSIT, at zero spectral cost.

Proposed Precoder Design

• ZF without CSIT (baseline):

- Transmit: $\mathbf{x} = \mathbf{G}_* \mathbf{d} \Rightarrow \mathbf{r} = \mathbf{H} \mathbf{G}_* \mathbf{d} + \mathbf{n}$.
- ZF detection: $\hat{\mathbf{d}} = \text{DEC}\{\mathbf{G}_*^H \mathbf{H}^{-1} \mathbf{r}\}$, $\mathbf{w} \sim \mathcal{CN}(\mathbf{0}, \sigma^2 \mathbf{G}_*^H (\mathbf{H}^H \mathbf{H})^{-1} \mathbf{G}_*)$.

• ZF with CSIT:

- Apply SVD: $\mathbf{H} \mathbf{G}_* = \mathbf{U}_c \mathbf{\Gamma}_c \mathbf{V}_c^H$, $\mathbf{\Gamma}_c = \text{diag}\{\gamma_1, \dots, \gamma_D\}$.
- Transmit: $\mathbf{x} = \mathbf{G}_* \mathbf{V}_c \mathbf{d} \Rightarrow \mathbf{r} = \mathbf{U}_c \mathbf{\Gamma}_c \mathbf{d} + \mathbf{n}$.
- ZF detection: $\hat{\mathbf{d}} = \text{DEC}\{\mathbf{\Gamma}_c^{-1} \mathbf{U}_c^H \mathbf{r}\}$, $\mathbf{w}_c \sim \mathcal{CN}(\mathbf{0}, \sigma^2 \mathbf{\Gamma}_c^{-2})$.

• SIC with CSIT:

- Apply GMD: $\mathbf{H} \mathbf{G}_* = \gamma \mathbf{Q}_c \mathbf{R}_c \mathbf{P}_c^H$, $\gamma = \left(\prod_{i=1}^D \gamma_i\right)^{1/D}$, where $\mathbf{R}_c$ is unit upper triangular and $\mathbf{Q}_c, \mathbf{P}_c$ (semi)-unitary.
- Transmit: $\mathbf{x} = \mathbf{G}_* \mathbf{P}_c \mathbf{d}$.
- After projection: $\tilde{\mathbf{r}} \triangleq \gamma^{-1} \mathbf{Q}_c^H \mathbf{r} = \mathbf{R}_c \mathbf{d} + \tilde{\mathbf{w}}_c$, where $\tilde{\mathbf{w}}_c \sim \mathcal{CN}(\mathbf{0}, \frac{\sigma^2}{\gamma^2} \mathbf{I}_D)$.
- Backward SIC:

$$\hat{d}_D = \text{DEC}\{\tilde{r}_D\}, \quad \hat{d}_\tau = \text{DEC}\{\tilde{r}_\tau - \sum_{\nu > \tau} \rho_{\tau\nu} \hat{d}_\nu\}, \quad \tau = D-1, \dots, 1.$$

Numerical Results

• Simulation setup:

- **Waveform & channel:** CP-OFDM, rect. pulse, ideal LP DAC; $N = 256$, $l_g = 32$, $K = 129$ contiguous tones; flat $W(f)$ on $\mathcal{B}$; block-fading exp. PDP factor $\delta = 0.1$, $\sum_{k \in \mathcal{K}} |H_c[k]|^2 = K$, 16-QAM modulation.

• Sidelobe suppression:

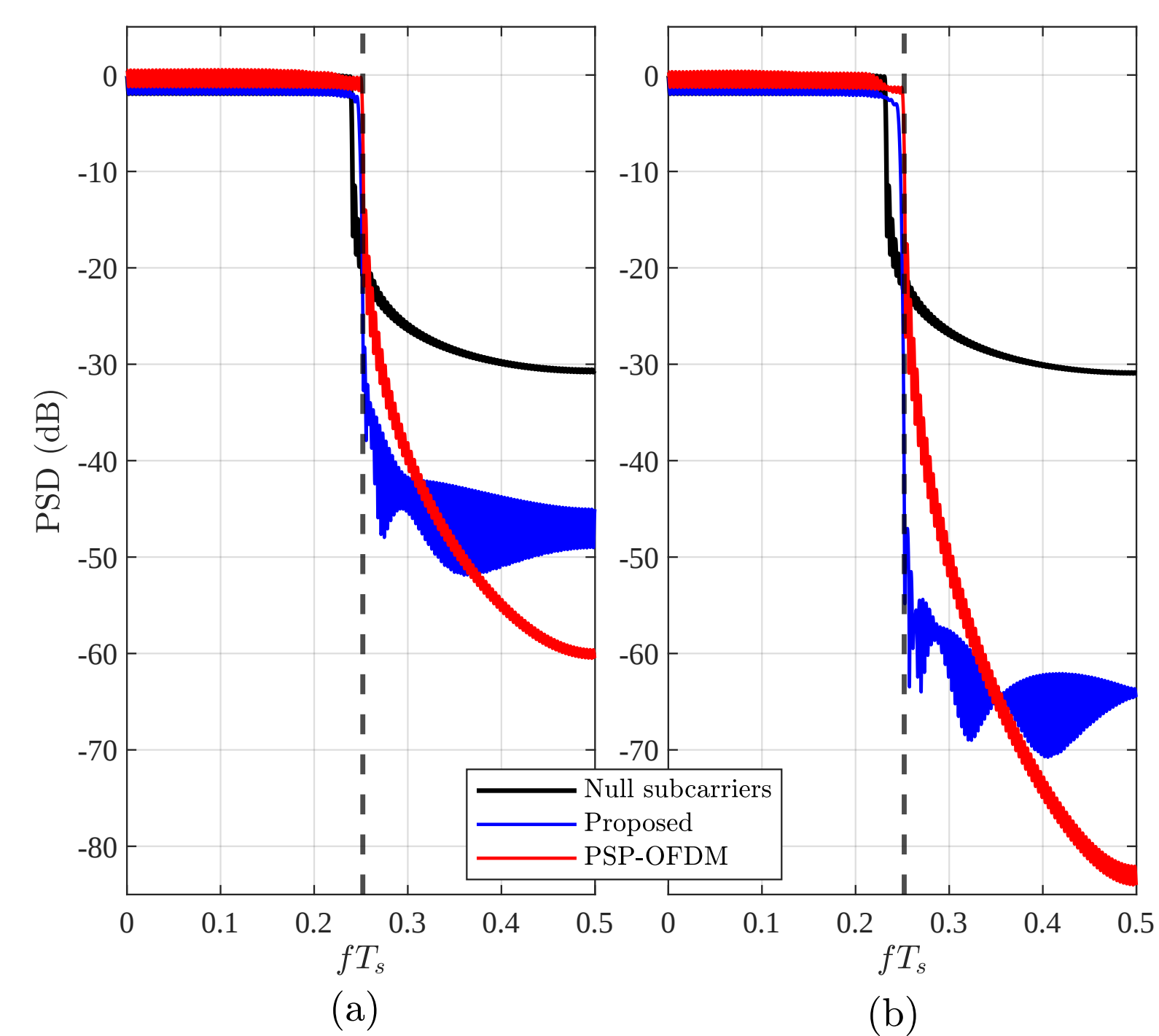


Fig. 2: PSDs of proposed and PSP-OFDM [2] precoders ($\delta = 0.10$); with $R = 6$ (left), and $R = 10$ (right).

• Error rate:

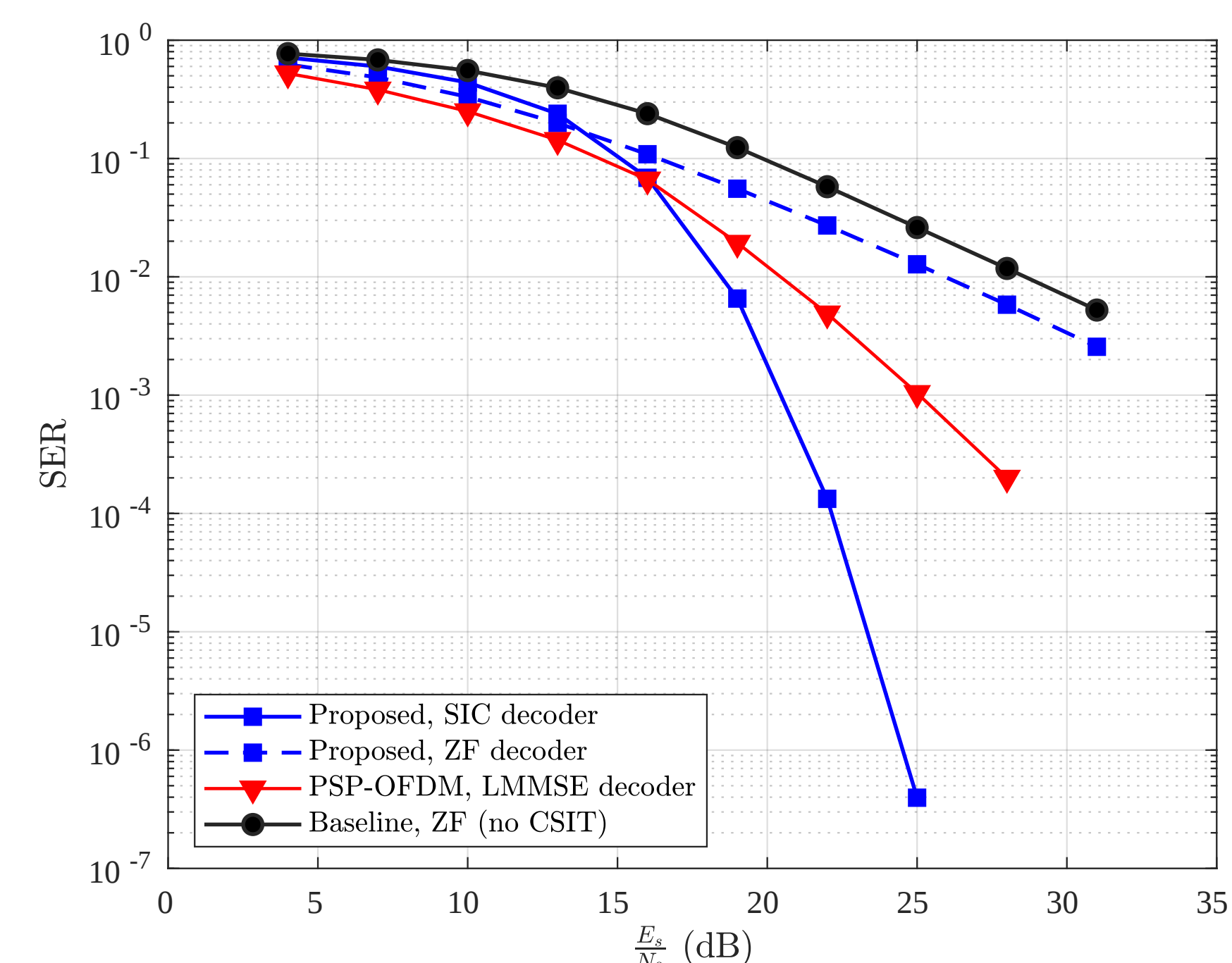


Fig. 3: SER performance of proposed and PSP-OFDM [2] precoders (ZF and SIC), with $\delta = 0.10$ and $R = 6$.

References & Acknowledgment

- [1] R. López-Valcarce, "General form of the power spectral density of multicarrier signals," *IEEE Commun. Lett.*, vol. 26, no. 8, pp. 1755–1759, 2022.
- [2] W.-C. Chen, C.-D. Chung, and P.-H. Wang, "Pre-equalized and spectrally precoded OFDM," *IEEE Trans. Veh. Technol.*, vol. 71, no. 7, pp. 7472–7486, 2022.

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