



UNIVERSITAT POLITÈCNICA  
DE CATALUNYA  
BARCELONATECH

# Communication Rates for Fading Channels with Imperfect Channel-State Information

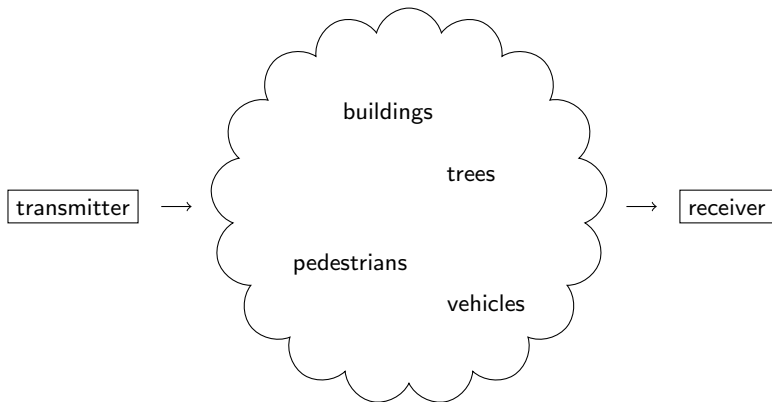
PhD Defense

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Universitat Politècnica de Catalunya  
Dep. of Signal Theory and Communications  
Barcelona, Spain

June 27, 2014

## What is fading?



The communication suffers from:

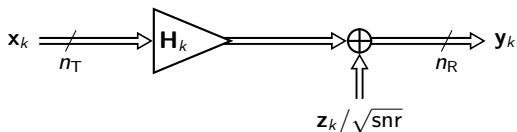
- shadowing
- scattering
- multiple reflections

all of which **vary in time!**

# Outline

- ① Improved Capacity Bounds for Imperfect CSI
- ② Joint Pilot-Precoder Design

## MIMO system with perfect CSI

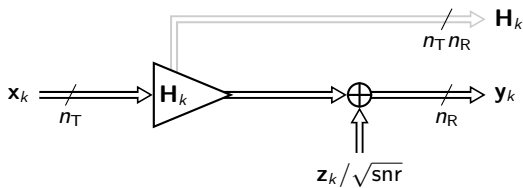


## Discrete-time system equation

At time instant  $k$ , the output  $\mathbf{y}_k$  corresponding to the input  $\mathbf{x}_k$  is

$$\mathbf{y}_k = \mathbf{H}_k \mathbf{x}_k + \frac{1}{\sqrt{\text{snr}}} \mathbf{z}_k \quad (k \in \mathbb{Z})$$

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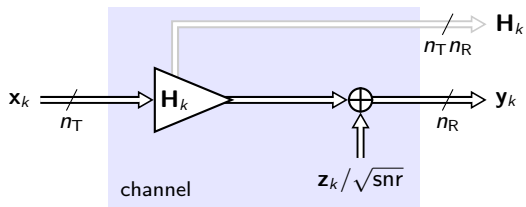


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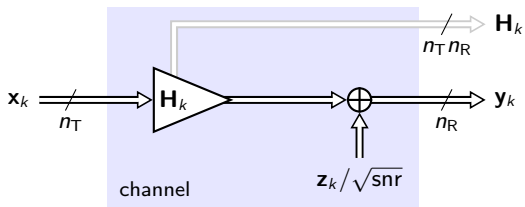


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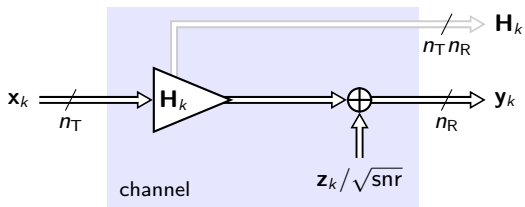
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**Assumptions:**

- $\{\mathbf{H}_k\}$ ,  $\{\mathbf{x}_k\}$  and  $\{\mathbf{z}_k\}$  are mutually independent i.i.d. sequences
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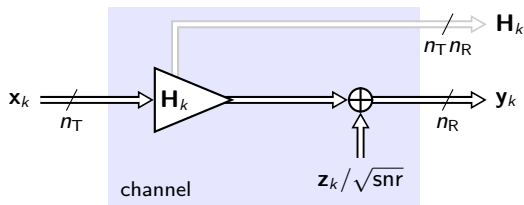
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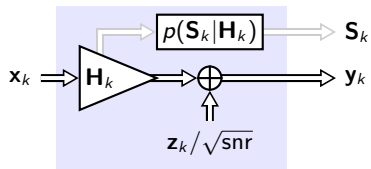
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**Coding theorem**

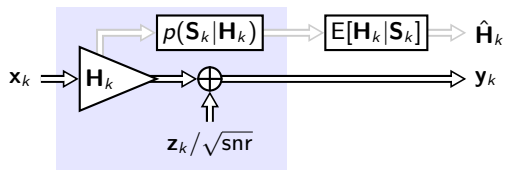
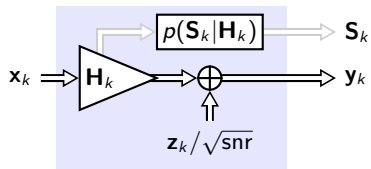
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is an achievable rate.

## MIMO system with imperfect CSI (1 of 2)

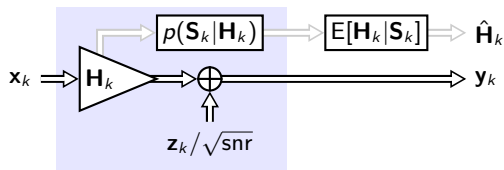
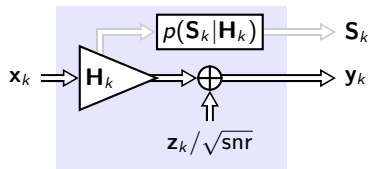


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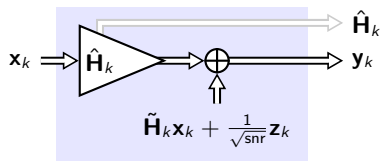


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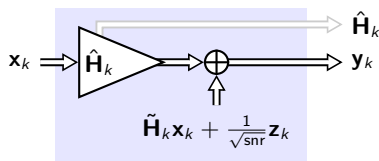


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Channel error:  
 $\tilde{\mathbf{H}}_k = \mathbf{H}_k - \hat{\mathbf{H}}_k$   
 $E[\tilde{\mathbf{H}}_k|\hat{\mathbf{H}}_k] = \mathbf{0}$

## MIMO system with imperfect CSI (2 of 2)



## Discrete-time system equation

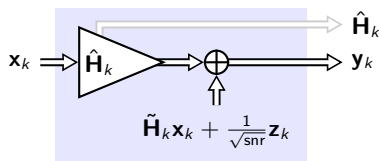
At time instant  $k \in \mathbb{Z}$ , the system output is

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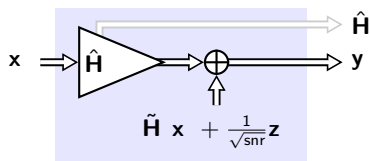
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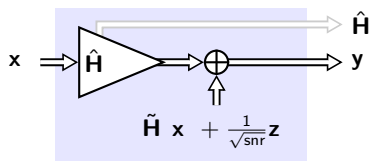
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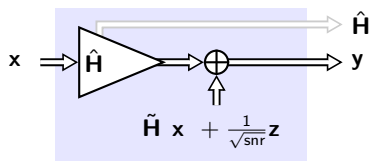
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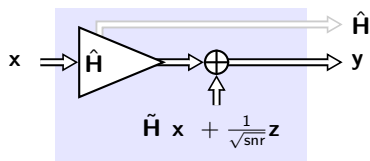
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### Worst-case noise bound

[Médard '00, Hassibi & Hochwald '03, Yoo & Goldsmith '06]

$$\begin{aligned} I(\mathbf{x}_G; \mathbf{y} | \hat{\mathbf{H}}) &\geq E_{\hat{\mathbf{H}}} \left[ \log \det \left( \mathbf{I} + \underbrace{\left( E[\tilde{\mathbf{H}}\mathbf{Q}\tilde{\mathbf{H}}^\dagger | \hat{\mathbf{H}}] + \frac{1}{\text{snr}} E[\mathbf{z}\mathbf{z}^\dagger] \right)}_{\text{effective noise covariance } \mathbf{\Gamma}_{\text{eff}}}^{-1} \hat{\mathbf{H}}\mathbf{Q}\hat{\mathbf{H}}^\dagger \right) \right] \\ &\triangleq R_{\text{WCN}} \end{aligned}$$

Assume  $E[\mathbf{z}\mathbf{z}^\dagger] = \mathbf{I}_r$  without loss of generality

## Covariance-constrained capacity

- Perfect CSIR [Foschini & Gans '98, Telatar '99]:

$$C_{\text{coh}} = \sup_{\mathbb{E}[\mathbf{x}\mathbf{x}^\dagger] = \mathbf{Q}} I(\mathbf{x}; \mathbf{y} | \mathbf{H}) = I(\mathbf{x}_G; \mathbf{y} | \mathbf{H}) = \mathbb{E}_{\mathbf{H}} \left[ \log \det(\mathbf{I} + \mathbf{\Gamma}^{-1} \mathbf{H} \mathbf{Q} \mathbf{H}^\dagger) \right]$$

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**Can we improve the lower bound?**

$$C \geq I(\mathbf{x}_G; \mathbf{y}|\hat{\mathbf{H}}) \geq \mathbf{E} \left[ \log \det(\mathbf{I} + \mathbf{\Gamma}_{\text{eff}}^{-1} \hat{\mathbf{H}} \mathbf{Q} \hat{\mathbf{H}}^\dagger) \right]$$



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## Rate splitting: How does it help? (1 of 3)

Decompose  $\mathbf{x}_G$  into a sum of two independent Gaussian signals:

$$\underbrace{\mathbf{x}_G}_{\mathcal{N}_{\mathbb{C}}(\mathbf{0}, \mathbf{Q})} = \underbrace{\mathbf{x}_1}_{\mathcal{N}_{\mathbb{C}}(\mathbf{0}, \mathbf{Q}_1)} + \underbrace{\mathbf{x}_2}_{\mathcal{N}_{\mathbb{C}}(\mathbf{0}, \mathbf{Q}_2)} \quad (\mathbf{Q} = \mathbf{Q}_1 + \mathbf{Q}_2)$$

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### Chain rule

$$\begin{aligned} I(\mathbf{x}_G; \mathbf{y} | \hat{\mathbf{H}}) &= I(\mathbf{x}_1, \mathbf{x}_2; \mathbf{y} | \hat{\mathbf{H}}) \\ &= I(\mathbf{x}_1; \mathbf{y} | \hat{\mathbf{H}}) + I(\mathbf{x}_2; \mathbf{y} | \hat{\mathbf{H}}, \mathbf{x}_1) \end{aligned}$$

## Rate splitting: How does it help? (2 of 3)

1) First term  $I(\mathbf{x}_1; \mathbf{y} | \hat{\mathbf{H}}) \geq R_{\text{WCN},1}$ :

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2) Second term  $I(\mathbf{x}_2; \mathbf{y} | \hat{\mathbf{H}}, \mathbf{x}_1) \geq R_{\text{WCN},2}$ : (“decode  $\mathbf{x}_2$  knowing  $\hat{\mathbf{H}}$  and  $\mathbf{x}_1$ ”)

## Rate splitting: How does it help? (2 of 3)

1) First term  $I(\mathbf{x}_1; \mathbf{y} | \hat{\mathbf{H}}) \geq R_{\text{WCN},1}$ : (“decode  $\mathbf{x}_1$  knowing  $\hat{\mathbf{H}}$ ”)

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## Rate splitting: How does it help? (3 of 3)

$$\begin{array}{rcl}
 \text{WCN bound:} & R_{\text{WCN}} & \leq I(\mathbf{x}_G; \mathbf{y} | \hat{\mathbf{H}}) \\
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 \end{array}
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$$\stackrel{\text{Jensen's ineq.}}{\geq} E_{\hat{\mathbf{H}}} \left[ \log \det \left( \mathbf{I} + (\dots \mathbf{Q}_1 \dots)^{-1} \hat{\mathbf{H}} \mathbf{Q}_2 \hat{\mathbf{H}}^\dagger \right) \right]$$

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## Rate splitting: How does it help? (3 of 3)

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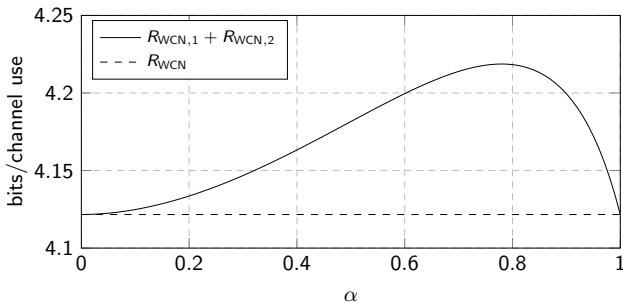
$$\geq E_{\hat{\mathbf{H}}} \left[ \log \det \left( \mathbf{I} + (\dots \mathbf{Q}_1 \dots)^{-1} \hat{\mathbf{H}} \mathbf{Q}_2 \hat{\mathbf{H}}^\dagger \right) \right]$$

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Rate splitting helps!

$$R_{\text{WCN}} \leq R_{\text{WCN},1} + R_{\text{WCN},2} \triangleq R_{\text{RS}}$$

## Rate splitting: numerical example



Rate-splitting bound  $R_{WCN,1} + R_{WCN,2}$  compared to  $R_{WCN}$  for  $(\mathbf{Q}_1, \mathbf{Q}_2) = (\alpha\mathbf{Q}, (1-\alpha)\mathbf{Q})$  and  $\text{snr} = 10 \text{ dB}$ ,  $n_R = 6$ ,  $n_T = 4$

## Rate splitting: general formula

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- $(\mathbf{Q}_1, \mathbf{Q}_2) = (\alpha\mathbf{Q}, (1 - \alpha)\mathbf{Q})$  is not the only possible decomposition!  
For example:

$$\mathbf{Q}_1 = \lambda_1 \mathbf{u}_1 \mathbf{u}_1^\dagger$$

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- How about more than 2 splits?

$$\mathbf{Q} = \mathbf{Q}_1 + \mathbf{Q}_2 + \dots + \mathbf{Q}_L$$

## Rate splitting: general formula

Decompose  $\mathbf{x}_G$  into a sum of  $L$  signals (layers):

$$\underbrace{\mathbf{x}_G}_{\mathcal{N}_{\mathbf{C}}(\mathbf{0}, \mathbf{Q})} = \sum_{\ell=1}^L \underbrace{\mathbf{x}_\ell}_{\mathcal{N}_{\mathbf{C}}(\mathbf{0}, \mathbf{Q}_\ell)} \quad \mathbf{Q} = \sum_{\ell=1}^L \mathbf{Q}_\ell$$

Worst-case noise bound with  $L$ -layer rate splitting

$$R_{\text{WCN}} \stackrel{\text{Jensen}}{\leq} \underbrace{\sum_{\ell=1}^L \mathbb{E} \log \det \left( \mathbf{I} + \hat{\mathbf{H}}^\dagger \Gamma_\ell^{-1} \hat{\mathbf{H}} \mathbf{Q}_\ell \right)}_{R_{\text{RS}}(\mathbf{Q}_1, \dots, \mathbf{Q}_L)} \stackrel{\text{chain rule} + \text{WCN}}{\leq} I(\mathbf{x}_G; \mathbf{y} | \hat{\mathbf{H}})$$

with

$$\Gamma_\ell = \underbrace{\mathbb{E} \left[ \tilde{\mathbf{H}} \left( \sum_{\ell'=1}^{\ell-1} \mathbf{x}_{\ell'} \right) \left( \sum_{\ell'=1}^{\ell-1} \mathbf{x}_{\ell'} \right)^\dagger \tilde{\mathbf{H}}^\dagger \mid \hat{\mathbf{H}}, \mathbf{x}_1, \dots, \mathbf{x}_{\ell-1} \right]}_{\text{residual interference from decoded layers}} + \underbrace{\mathbb{E} \left[ \tilde{\mathbf{H}} \mathbf{Q}_\ell \tilde{\mathbf{H}}^\dagger \mid \hat{\mathbf{H}} \right]}_{\text{CSIR imperfection for layer } \ell}$$

$$+ \underbrace{\hat{\mathbf{H}} \left( \mathbf{Q} - \sum_{\ell'=1}^{\ell} \mathbf{Q}_{\ell'} \right) \hat{\mathbf{H}}^\dagger + \mathbb{E} \left[ \tilde{\mathbf{H}} \left( \mathbf{Q} - \sum_{\ell'=1}^{\ell} \mathbf{Q}_{\ell'} \right) \tilde{\mathbf{H}}^\dagger \mid \hat{\mathbf{H}} \right]}_{\text{interference from layers yet to be decoded}} + \underbrace{\frac{1}{\text{snr}} \mathbf{I}}_{\text{additive noise}}$$



## Towards the best rate-splitting allocation (1 of 3)

- Riemann integration:

$$\sum_{i=1}^n f(x_i) \Delta x_i \xrightarrow{n \rightarrow \infty} \int_{x_0}^{x_\infty} f(x) dx$$

- Riemann-Stieltjes integration:

$$\sum_{i=1}^n f(x_i) g(\Delta x_i) \xrightarrow{n \rightarrow \infty} \int_{x_0}^{x_\infty} f(x) dg(x)$$

## Towards the best rate-splitting allocation (2 of 3)

## Layering functions

A layering function  $\mathbf{K}: [0; 1] \rightarrow \mathbb{C}_+^{n_T \times n_T}$  has the following properties:

- continuity
- $0 \leq \lambda_1 < \lambda_2 \leq 1 \Rightarrow \mathbf{K}(\lambda_1) \stackrel{\neq}{\preceq} \mathbf{K}(\lambda_2)$
- $\mathbf{K}(0) = \mathbf{0}$
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Given a pair  $(\mathbf{K}, L)$ , a tuple  $(\mathbf{Q}_1, \dots, \mathbf{Q}_L)$  is uniquely determined via

$$\mathbf{Q}_\ell = \mathbf{K}\left(\frac{\ell}{L}\right) - \mathbf{K}\left(\frac{\ell-1}{L}\right), \quad \ell = 1, \dots, L$$

such that

$$R_{\text{RS}}(\mathbf{Q}_1, \dots, \mathbf{Q}_L) \triangleq R_{\text{RS}}(\mathbf{K}, L)$$

## Towards the best rate-splitting allocation (3 of 3)

The best rate-splitting bound is

$$R_{RS}^{**} = \sup_L \sup_{\substack{\mathbf{Q}_1, \dots, \mathbf{Q}_L \\ \sum_{\ell=1}^L \mathbf{Q}_\ell = \mathbf{Q}}} R_{RS}(\mathbf{Q}_1, \dots, \mathbf{Q}_L)$$

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## Theorem ("Infinite" layering)

$R_{RS}^*(\mathbf{K})$  is achieved for  $L \rightarrow \infty$  and given by the Riemann-Stieltjes integral

$$R_{RS}^*(\mathbf{K}) = \lim_{L \rightarrow \infty} R_{RS}(\mathbf{K}, L) = \int_0^1 \mathbb{E} \left[ \text{tr} \left\{ \hat{\mathbf{H}}^\dagger \Gamma(\lambda)^{-1} \hat{\mathbf{H}} d\mathbf{K}(\lambda) \right\} \right]$$

with  $\boldsymbol{\xi} \sim \mathcal{N}_{\mathbb{C}}(\mathbf{0}, \mathbf{I})$  and  $\Gamma(\lambda) =$

$$\underbrace{\mathbb{E} \left[ \tilde{\mathbf{H}} \mathbf{K}(\lambda)^{\frac{1}{2}} \boldsymbol{\xi} \boldsymbol{\xi}^\dagger \mathbf{K}(\lambda)^{\frac{1}{2}} \tilde{\mathbf{H}}^\dagger \mid \boldsymbol{\xi} \right]}_{\text{residual interference from decoded layers}} + \underbrace{\hat{\mathbf{H}}(\mathbf{Q} - \mathbf{K}(\lambda)) \hat{\mathbf{H}}^\dagger + \mathbb{E} \left[ \tilde{\mathbf{H}}(\mathbf{Q} - \mathbf{K}(\lambda)) \tilde{\mathbf{H}}^\dagger \right]}_{\text{interference from layers yet to be decoded}} + \underbrace{\frac{1}{\text{snr}} \mathbf{I}_{n_R}}_{\text{additive noise}}$$

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## Remarks:

- Compact notation for  $\int_0^1 \text{tr} [\mathbf{A}(\lambda) d\mathbf{B}(\lambda)] = \sum_{i,j} \int_0^1 A_{i,j}(\lambda) dB_{j,i}(\lambda)$
- For continuously differentiable  $\mathbf{K}$ , we have  $d\mathbf{K}(\lambda) = \mathbf{K}'(\lambda) d\lambda$  and we obtain a Riemann integral
- $R_{\text{WCN}}$  can be recovered via Jensen's inequality

## Proof sketch of the infinite-layering formula

**First step (“adding layers always helps”):**

$$R_{\text{RS}}(\dots, \mathbf{Q}_\ell + \mathbf{Q}_{\ell+1}, \dots) \leq R_{\text{RS}}(\dots, \mathbf{Q}_\ell, \mathbf{Q}_{\ell+1}, \dots) \Rightarrow R_{\text{RS}}^*(\mathbf{K}) = \lim_{L \rightarrow \infty} R_{\text{RS}}(\mathbf{K}, L)$$

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Denote  $\Delta \mathbf{K}_\ell = \mathbf{K}(\frac{\ell}{L}) - \mathbf{K}(\frac{\ell-1}{L}) = \mathbf{Q}_\ell$ .

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**Second step (“from sum to integral”):**

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$$\begin{aligned} R_{\text{RS}}(\mathbf{K}, L) &= \sum_{\ell=1}^L \mathbb{E} \left[ \log \det \left( \mathbf{I} + \hat{\mathbf{H}}^\dagger \mathbf{\Gamma}_\ell^{-1} \hat{\mathbf{H}} \Delta \mathbf{K}_\ell \right) \right] \\ &= \sum_{\ell=1}^L \left[ \mathbb{E} \left[ \text{tr} \left\{ \hat{\mathbf{H}}^\dagger \mathbf{\Gamma}_\ell^{-1} \hat{\mathbf{H}} \Delta \mathbf{K}_\ell \right\} \right] + \mathcal{O}(\Delta \mathbf{K}_\ell^2) \right] \\ &\xrightarrow{L \rightarrow \infty} \boxed{\int_0^1 \mathbb{E} \left[ \text{tr} \left\{ \hat{\mathbf{H}}^\dagger \mathbf{\Gamma}(\lambda)^{-1} \hat{\mathbf{H}} d\mathbf{K}(\lambda) \right\} \right]} + \underbrace{\int_0^1 d\lambda^2}_{=0} \end{aligned}$$

## Three additional properties of layering functions

### Continuity

Provided that  $E[\|\hat{\mathbf{H}}\|_F^4] < \infty$ , the function  $R_{RS}^*(\mathbf{K})$  is uniformly continuous in  $\mathbf{K}$ :

$$\forall \epsilon > 0: \exists \delta > 0: \|\mathbf{K}_1 - \mathbf{K}_2\|_\infty \leq \delta \Rightarrow |R_{RS}^*(\mathbf{K}_1) - R_{RS}^*(\mathbf{K}_2)| \leq \epsilon$$



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### Optimal layering

Provided that  $E[\|\hat{\mathbf{H}}\|_F^4] < \infty$ , there exists an optimal layering function

$$\mathbf{K}^* = \underset{\mathbf{K}}{\operatorname{argmax}} R_{RS}^*(\mathbf{K})$$

such that  $R_{RS}^{**} = R_{RS}^*(\mathbf{K}^*)$ .

## Example 1: SIMO with white channel error

Assumptions:

- $n_T = 1$ ,  $(\hat{\mathbf{H}}, \tilde{\mathbf{H}}) \rightarrow (\hat{\mathbf{h}}, \tilde{\mathbf{h}})$ ,  $\mathbf{K}(\lambda) \rightarrow k(\lambda)$
- $\hat{\mathbf{h}}$  and  $\tilde{\mathbf{h}}$  are mutually independent
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After some algebra:

$$R_{RS}^*(k) = \int_0^1 \mathbb{E} \left[ \frac{\|\hat{\mathbf{h}}\|_2^2}{1 - k(\lambda) + \beta(k(\lambda))\|\hat{\mathbf{h}}\|_2^2} \right] dk(\lambda)$$

where  $\beta(k) = \tilde{\mathbf{V}}((\Xi_1 - 1)k + 1) + \text{snr}^{-1}$  with  $\Xi_1 \sim \text{gamma}(1, 1)$

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$$R_{\text{RS}}^* = \mathbb{E} \left[ \frac{\|\hat{\mathbf{h}}\|_2^2}{\tilde{\mathbf{V}}(\Xi_1 - 1) - \|\hat{\mathbf{h}}\|_2^2} \log \left( 1 + \frac{\tilde{\mathbf{V}}(\Xi_1 - 1) - \|\hat{\mathbf{h}}\|_2^2}{\|\hat{\mathbf{h}}\|_2^2 + \tilde{\mathbf{V}} + \text{snr}^{-1}} \right) \right]$$

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**Layering function has disappeared!**



## Example 2: MIMO with i.i.d. fading and two exemplary layerings (1 of 2)

Assumptions:

- $\text{vec}(\hat{\mathbf{H}}) \sim \mathcal{N}_{\mathbb{C}}(\mathbf{0}, \hat{\mathbf{V}}\mathbf{I}_{n_{\text{R}}n_{\text{T}}})$  and  $\text{vec}(\tilde{\mathbf{H}}) \sim \mathcal{N}_{\mathbb{C}}(\mathbf{0}, \tilde{\mathbf{V}}\mathbf{I}_{n_{\text{R}}n_{\text{T}}})$  are independent
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Levelled layering  $\mathbf{K}_{\text{lev}}$ 

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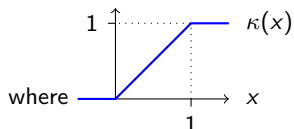
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### Staggered layering $\mathbf{K}_{\text{stag}}$

$$\mathbf{K}_{\text{stag}}(\lambda) = \frac{1}{n_{\text{T}}} \begin{bmatrix} \kappa(\lambda n_{\text{T}}) & & & & & & 0 \\ & \kappa(\lambda n_{\text{T}} - 1) & & & & & \\ & & & \ddots & & & \\ & & & & \ddots & & \\ 0 & & & & & \kappa(\lambda n_{\text{T}} - n_{\text{T}} + 1) & \end{bmatrix}$$



## Example 2: MIMO with i.i.d. fading and two exemplary layerings (2 of 2)

- **Levelled layering:**

$$R_{RS}^*(\mathbf{K}_{\text{lev}}) = \frac{1}{n_T} \int_0^1 \mathbb{E} \left[ \text{tr} \left\{ \hat{\mathbf{H}}^\dagger \mathbf{I}_{\text{lev}}(\lambda)^{-1} \hat{\mathbf{H}} \right\} \right] d\lambda$$

where

$$\mathbf{I}_{\text{lev}}(\lambda) = \left( \lambda \tilde{V} \frac{\Xi_{n_T}}{n_T} + \tilde{V}(1-\lambda) + \frac{1}{\text{snr}} \right) \mathbf{I}_{n_R} + (1-\lambda) \frac{1}{n_T} \hat{\mathbf{H}} \hat{\mathbf{H}}^\dagger$$

and  $\Xi_{n_T} \sim \text{gamma}(n_T, 1)$

- **Staggered layering:**

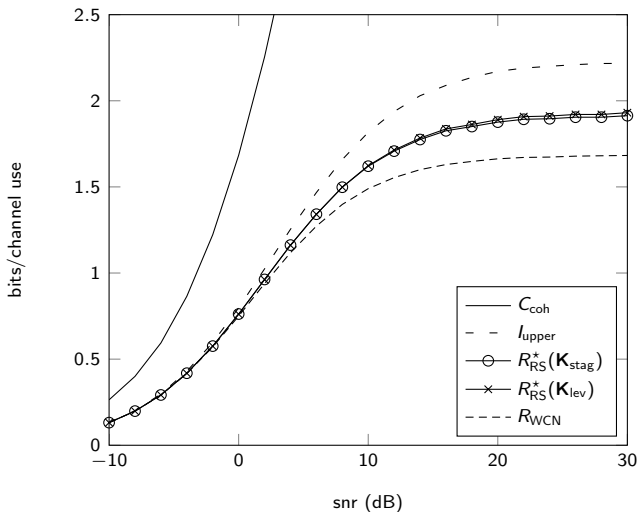
$$R_{RS}^*(\mathbf{K}_{\text{stag}}) = \sum_{i=1}^{n_T} \int_0^1 \mathbb{E} \left[ \frac{\hat{\mathbf{h}}_i^\dagger \mathbf{A}_i(\lambda)^{-1} \hat{\mathbf{h}}_i}{1 + (1-\lambda) \hat{\mathbf{h}}_i^\dagger \mathbf{A}_i(\lambda)^{-1} \hat{\mathbf{h}}_i} \right] d\lambda$$

where

$$\mathbf{A}_i(\lambda) = \left( \tilde{V} \left( \Xi_{i-1}^{(1)} + \lambda \Xi_1^{(2)} + 1 - \lambda + n_T - i \right) + \frac{n_T}{\text{snr}} \right) \mathbf{I}_{n_R} + \hat{\mathbf{H}}_{(i+1):n_T} \hat{\mathbf{H}}_{(i+1):n_T}^\dagger$$

and the  $\Xi_i^{(j)} \sim \text{gamma}(i, 1)$  are independent

## Numerical example



Capacity and MI bounds for a  $2 \times 2$  channel with  $\hat{\mathbf{H}} \sim \mathcal{N}_{\mathbb{C}}(\mathbf{0}, \frac{1}{2} \mathbf{I}_{n_R n_T})$  and  $\tilde{\mathbf{H}} \sim \mathcal{N}_{\mathbb{C}}(\mathbf{0}, \frac{1}{2} \mathbf{I}_{n_R n_T})$

## Large antenna arrays (1 of 2)

Assumptions:

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### Large transmit array ( $n_T \rightarrow \infty$ )

Consider a sequence  $(\mathbf{K}_{n_T})_{n_T \in \mathbb{N}}$  of  $n_T \times n_T$  layering functions. Then

$$\lim_{n_T \rightarrow \infty} \left\{ R_{\text{RS}}^*(\mathbf{K}_{n_T}) - R_{\text{WCN}} \right\} = 0$$

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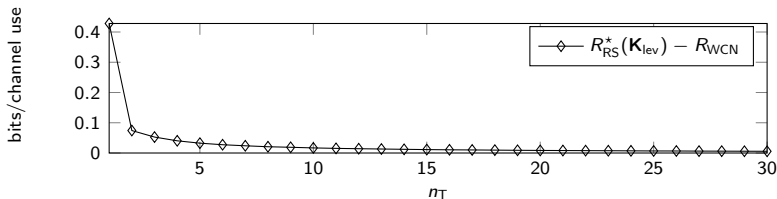
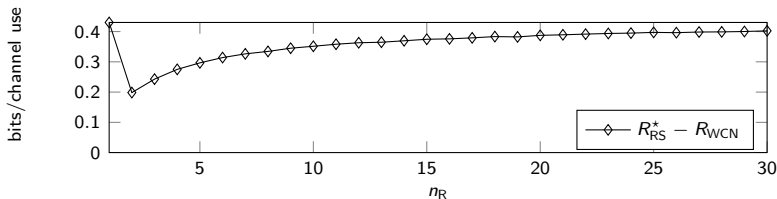
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## Large antenna arrays (2 of 2)

Bound difference  $R_{RS}^*(\mathbf{K}_{lev}) - R_{WCN}$  for a MISO channel ( $n_R = 1$ )Bound difference  $R_{RS}^* - R_{WCN}$  for a SIMO channel ( $n_T = 1$ )

## Asymptotically perfect CSI

### Assumptions:

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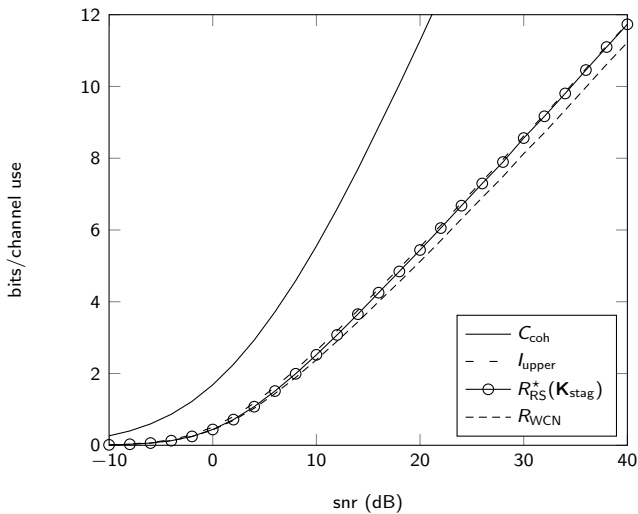
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### Asymptotic tightness

If  $\lim_{\text{snr} \rightarrow \infty} \tilde{V}_{\text{snr}} = 0$ , then  $R_{\text{RS}}^*(\mathbf{K}_{\text{snr}})$  is asymptotically tight:

$$\lim_{\text{snr} \rightarrow \infty} \left\{ I(\mathbf{x}_G; \mathbf{y} | \hat{\mathbf{H}}) - R_{\text{RS}}^*(\mathbf{K}_{\text{snr}}) \right\} = 0$$

## Asymptotically perfect CSI: numerical example



Capacity and MI bounds for a  $2 \times 2$  channel with  $\tilde{V}_{\text{snr}} = (1 + \text{snr})^{-1/2}$



## Introduction

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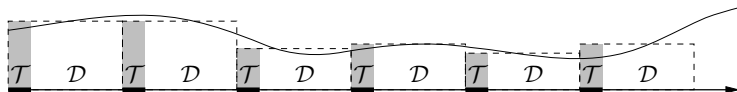
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## Introduction

- Pilot-aided communication achieves good performance-complexity tradeoff
- Forward pilot symbols transmitted periodically
- Imperfect CSIR for the decoding task
- Variables to be optimized: **linear precoder** and **training sequence**

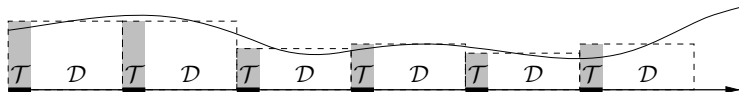
## Rayleigh block fading

Block-fading with training preambles:



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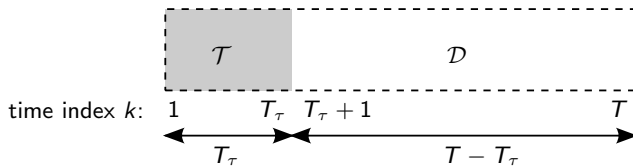
Transmit-side spatial correlation:

$$\mathbf{H} = \mathbf{W}\mathbf{R}^{\frac{1}{2}}$$

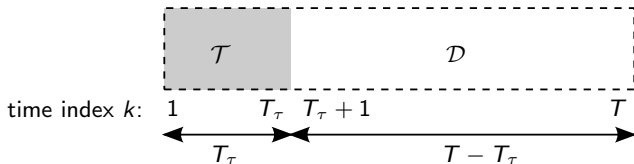
where

- $[\mathbf{W}]_{i,j}$  are random i.i.d.  $\mathcal{N}_{\mathbb{C}}(0, 1)$
- $\mathbf{R} = \mathbb{E}[\mathbf{H}^{\dagger}\mathbf{H}]$  is the channel correlation

## System model



## System model



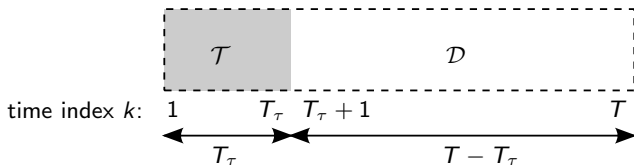
Training phase  $\mathcal{T}$ :

$$\mathbf{y}_\tau^{(k)} = \mathbf{H}\mathbf{t}^{(k)} + \mathbf{z}^{(k)}$$

Data transmission phase  $\mathcal{D}$ :

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Two matrix variables to be optimized:

- pilot Gram  $\mathbf{P} = [\mathbf{t}^{(1)} \dots \mathbf{t}^{(T_\tau)}]$   $\begin{bmatrix} \mathbf{t}^{(1)\dagger} \\ \vdots \\ \mathbf{t}^{(T_\tau)\dagger} \end{bmatrix}$
- $\text{tr}(\mathbf{P})$  is *training energy*
- $\text{rank}(\mathbf{P}) = T_\tau$  is *training duration*
- transmit covariance  $\mathbf{Q} = \mathbf{F}\mathbf{F}^\dagger$
- $\text{tr}(\mathbf{Q})$  is *transmit power*
- $\text{rank}(\mathbf{Q})$  is *number of streams*



## Worst-case noise lower bound

The WCN bound reads as

$$\begin{aligned}
 I(\mathbf{x}_G; \mathbf{y} | \hat{\mathbf{H}}) &\geq E_{\mathbf{W}} \left[ \log \det \left( \mathbf{I} + \mathbf{W} \underbrace{\frac{\mathbf{Q}^{\frac{1}{2}} (\mathbf{R} - (\mathbf{R}^{-1} + \mathbf{P})^{-1}) \mathbf{Q}^{\frac{1}{2}}}{1 + \text{tr}(\mathbf{Q}(\mathbf{R}^{-1} + \mathbf{P})^{-1})}}_{\mathbf{S}} \mathbf{W}^{\dagger} \right) \right] \\
 &= E_{\mathbf{W}} [\log \det(\mathbf{I} + \mathbf{W} \mathbf{S} \mathbf{W}^{\dagger})] \\
 &= R_{\text{WCN}}(\mathbf{S}) \\
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 \end{aligned}$$

and is a function of  $\mathbf{s} = \boldsymbol{\lambda}(\mathbf{S})$ .

## Utility functions I

## Definition

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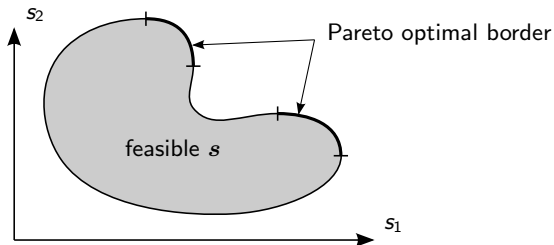
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Utility optimization comprises a multiobjective (Pareto) optimization:



## Utility functions II

Examples of utility functions:

- $f(\mathbf{S}) = \mathbb{E}[\log_2 \det(\mathbf{I} + \mathbf{W}\mathbf{S}\mathbf{W}^\dagger)]$
  - $f(\mathbf{S}) = -\text{tr} \mathbb{E}[(\mathbf{I} + \mathbf{W}\mathbf{S}\mathbf{W}^\dagger)^{-1}]$
  - $f(\mathbf{S}) = \lambda_{\min}(\mathbf{S})$
  - $f(\mathbf{S}) = \det(\mathbf{S})$
  - $f(\mathbf{S}) = \text{tr}(\mathbf{S}^{-1})^{-1}$
- $f(\mathbf{S}) = \Pr\{\det(\mathbf{I} + \mathbf{W}\mathbf{S}\mathbf{W}^\dagger) > \eta\}$
  - $f(\mathbf{S}) = \Pr\{-\text{tr}[(\mathbf{I} + \mathbf{W}\mathbf{S}\mathbf{W}^\dagger)^{-1}] > \eta\}$
  - $f(\mathbf{S}) = \lambda_{\max}(\mathbf{S})$
  - $f(\mathbf{S}) = \text{tr}(\mathbf{S})$
  - $f(\mathbf{S}) = \det(\mathbf{I} + \nu\mathbf{S})$

etc.

## Joint pilot/precoder optimization

$$(\mathbf{P}^*, \mathbf{Q}^*) = \underset{\substack{(\mathbf{P}, \mathbf{Q}) \\ \text{rank}(\mathbf{P})=T_\tau}}{\text{argmax}} \left\{ \frac{T - T_\tau}{T} f(\mathbf{S}(\mathbf{P}, \mathbf{Q})) \right\}$$

subject to:

- Positive semidefiniteness:  $\mathbf{P} \succeq \mathbf{0}$  and  $\mathbf{Q} \succeq \mathbf{0}$
- Energy conservation:  $\text{tr}(\mathbf{P}) + (T - T_\tau) \text{tr}(\mathbf{Q}) \leq T\mu$

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Reformulation:

$$(\mathbf{P}^*, \mathbf{Q}^*) = \underset{T_\tau=1, \dots, n_T}{\text{argmax}} \left\{ \frac{T - T_\tau}{T} \max_{\substack{(\mathbf{P}, \mathbf{Q}) \\ \text{rank}(\mathbf{P}) \leq T_\tau}} f(\mathbf{S}(\mathbf{P}, \mathbf{Q})) \right\}$$



## Partial optimization problems

- Prescribed precoder:  $\mathbf{P}^*(\mathbf{Q}) = \underset{\mathbf{P}}{\operatorname{argmax}} f(\mathbf{S}(\mathbf{P}, \mathbf{Q})) = \underset{\mathbf{S}(\bullet, \mathbf{Q})}{\operatorname{argmax}} f(\mathbf{S})$
- Prescribed pilots:  $\mathbf{Q}^*(\mathbf{P}) = \underset{\mathbf{Q}}{\operatorname{argmax}} f(\mathbf{S}(\mathbf{P}, \mathbf{Q})) = \underset{\mathbf{S}(\mathbf{P}, \bullet)}{\operatorname{argmax}} f(\mathbf{S})$

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### Theorem (rank inequalities)

$$\operatorname{rank}(\mathbf{P}^*(\mathbf{Q})) \leq \operatorname{rank}(\mathbf{Q})$$

$$\#pilots \leq \#streams$$

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### Theorem (convex reformulations)

*In the  $\mathbf{S}$ -domain, for (quasi-)concave utilities  $f$  the partial problems are (quasi-)convex!*

## Joint training and transmit directions

Spectral decompositions:

$$\mathbf{R} = \mathbf{U}_R \text{diag}(\mathbf{r}) \mathbf{U}_R^\dagger \quad \mathbf{P} = \mathbf{U}_P \text{diag}(\mathbf{p}) \mathbf{U}_P^\dagger \quad \mathbf{Q} = \mathbf{U}_Q \text{diag}(\mathbf{q}) \mathbf{U}_Q^\dagger$$

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### Theorem (eigenbasis alignment)

*For any utility function, it is optimal to set*

$$\mathbf{U}_P = \mathbf{U}_Q = \mathbf{U}_R$$

## Pareto optimization I

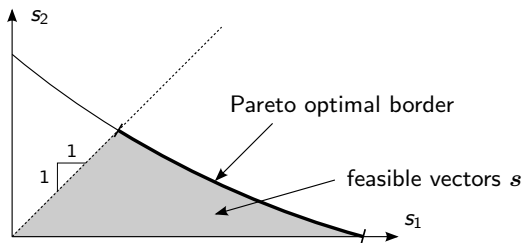
**Iterating between partial optimizations does not guarantee optimality!**

With alignment  $\mathbf{U}_P = \mathbf{U}_Q = \mathbf{U}_R$ , we get

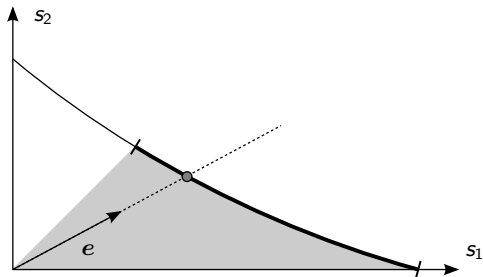
$$s_i = \frac{r_i^2 p_i q_i}{(1 + r_i p_i) \left(1 + \sum_{j=1}^{n_T} \frac{r_j q_j}{1 + r_j p_j}\right)}$$

Multiobjective (Pareto) optimization:

$$\max_{\substack{p_i, q_i \geq 0 \\ \sum_i p_i + (T - T_\tau) \sum_i q_i \leq T\mu}} (s_1, \dots, s_{n_T})$$



## Pareto optimization II



## Theorem (Pareto optimization)

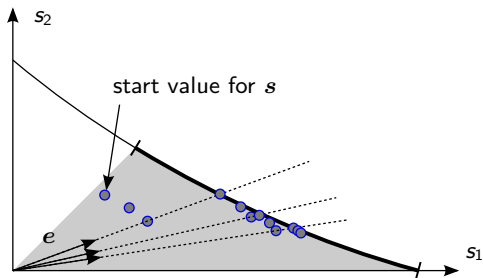
*The optimization along a fixed direction  $e$*

$$\max_{s/\|s\|=e} \|s\|$$

*is a **quasi-convex** problem*

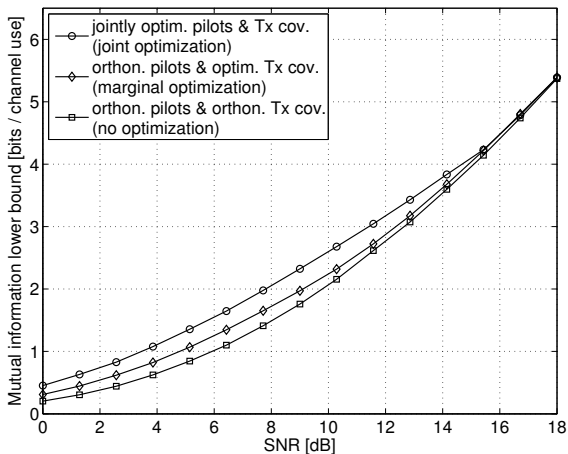
## Joint optimization algorithm

- ❶ Pilot optimization:  $\mathbf{s} \leftarrow \operatorname{argmax}_{\mathbf{s}(\bullet, \mathbf{Q})} f(\mathbf{s})$
- ❷ Precoder optimization:  $\mathbf{s} \leftarrow \operatorname{argmax}_{\mathbf{s}(\mathbf{P}, \bullet)} f(\mathbf{s})$
- ❸ Pareto optimization:  $\mathbf{s} \leftarrow \operatorname{argmax}_{\mathbf{s} / \|\mathbf{s}\| = e} \|\mathbf{s}\|$



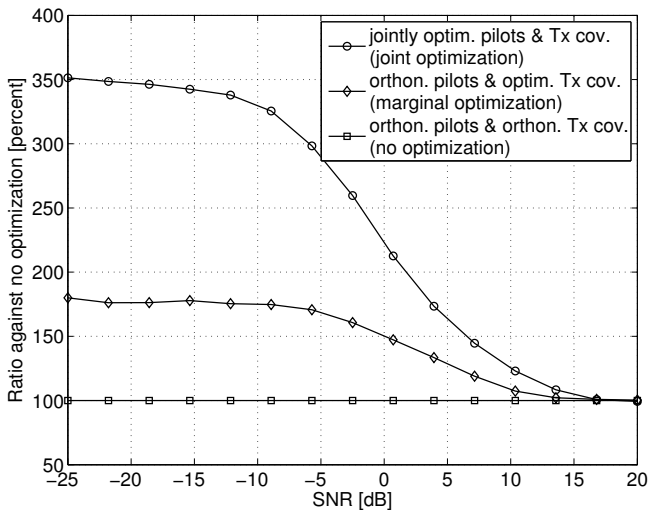


## Simulations I



2x2 MIMO system with  $r_1 = 0.7$ ,  $r_2 = 0.3$

## Simulations II



## Conclusions

### **Improved Capacity Bounds for Imperfect CSI:**

- novel bound on MIMO capacity with imperfect CSIR
- sharper yet a bit more complex
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Thank you!

Backup slides

## Large random-matrix approximation

Large random-matrix approximation [Hachem, Khorunzhiy, Loubaton, Najim, Pastur '08]

As  $n_R, n_T \rightarrow \infty$  with  $0 < \liminf(n_R/n_T) \leq \limsup(n_R/n_T) < \infty$ ,

$$R_{RS}^*(\mathbf{K}) = \bar{R}_{RS}^*(\mathbf{K}) + \mathcal{O}(1/n_T)$$

where

$$\bar{R}_{RS}^*(\mathbf{K}) = \int_0^\infty \int_{\sigma^2(x)}^\infty \frac{n_R - n_T g(x, u) d(g(x, u))}{u} f(x) du dx$$

with

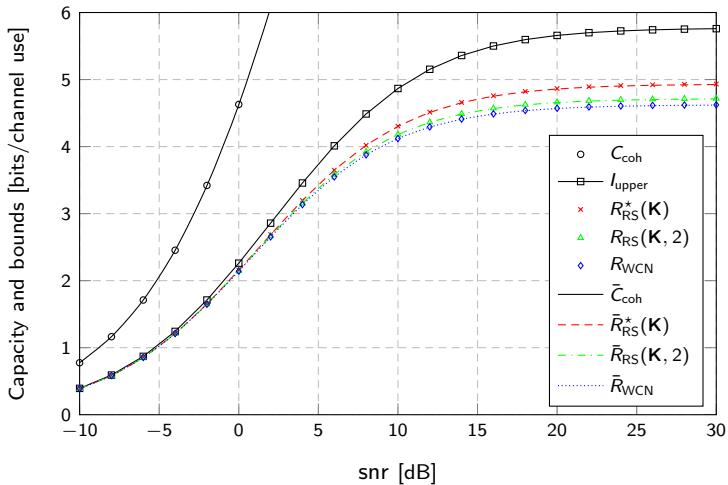
$$\sigma^2(x) = \frac{1}{\tilde{V}} \left( \frac{\tilde{V}}{n_T} x + \rho^{-1} \right)$$

$$f(x) = \frac{x^{n_T-1} e^{-x}}{(n_T - 1)!}$$

$$g(x, u) = \frac{\tilde{V}}{V} \left( 1 - \frac{x}{n_T} \right) + u$$

$$d(x) = \frac{\frac{n_R}{n_T} - 1}{2x} - \frac{1}{2} + \frac{\sqrt{\left(1 - \frac{n_R}{n_T} + x\right)^2 + 4 \frac{n_R}{n_T} x}}{2x}$$

## Large random-matrix approximation: numerical example



Capacity and MI bounds and their asymptotic approximations vs. SNR. Parameter values:  $n_{\text{R}} = 6$ ,  $n_{\text{T}} = 4$ ,  $\hat{\mathbf{V}} = \tilde{\mathbf{V}} = 0.5$ .



## Asymptotically perfect CSI (SISO)

### Assumptions:

- The joint law of  $(\hat{H}, \tilde{H}) = (\hat{H}_{\text{snr}}, \tilde{H}_{\text{snr}})$  depends on the SNR
- $\limsup_{\text{snr} \rightarrow \infty} \left\{ \frac{\tilde{V}_{\text{snr}}(\hat{h}_{\text{snr}})}{\frac{1}{\pi e} e^{h(\tilde{H}_{\text{snr}} | \hat{H}_{\text{snr}} = \hat{h}_{\text{snr}})}} \right\} \leq M$  for (almost) every  $\hat{H}_{\text{snr}} = \hat{h}_{\text{snr}}$

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### Asymptotic offset

If  $\lim_{\text{snr} \rightarrow \infty} \mathbb{E}[\tilde{V}_{\text{snr}}(\hat{H}_{\text{snr}})] = 0$ , then  $R_{\text{SD}}^*$  is asymptotically tight to within  $\log(M)$ :

$$\lim_{\text{snr} \rightarrow \infty} \left\{ I(X_G; Y | \hat{H}) - R_{\text{SD}}^* \right\} \leq \log(M)$$

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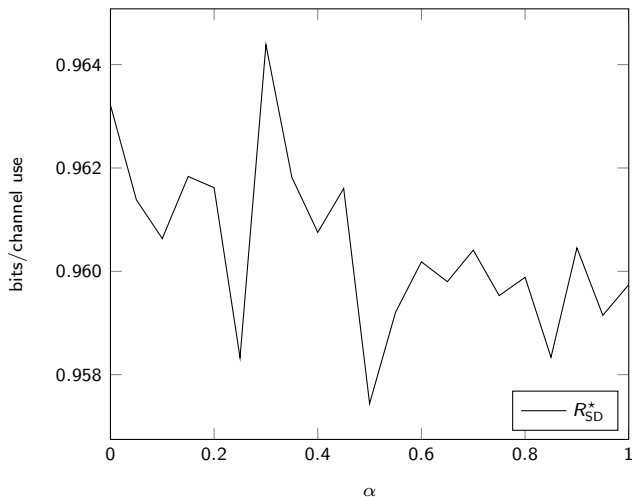
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### Corollary (asymptotic tightness)

If  $\tilde{H}_{\text{snr}} | \hat{H}_{\text{snr}}$  is Gaussian,

$$\lim_{\text{snr} \rightarrow \infty} \left\{ I(X_G; Y | \hat{H}) - R_{\text{SD}}^* \right\} = 0$$

## Dependency on the layering



Infinite-layer bound  $R_{SD}^*(\alpha\mathbf{K}_{lev} + (1 - \alpha)\mathbf{K}_{stag})$  as a function of  $\alpha$  for a  $(r, t) = (1, 2)$  MISO channel, snr = 13 dB